



5.5V, 2A Low R<sub>DSON</sub> Load Switch with Programmable Current Limit AEC-Q100 Qualified

## **DESCRIPTION**

The MPQ5073 is a load switch to provide 2A load protection covering 0.5V to 5.5V voltage range. With the small R<sub>DSON</sub> in tiny packages, MPQ5073 provides very high efficient and space saving solution in notebook and tablet or other portable devices application.

With the soft start function, the MPQ5073 can avoid inrush current during circuit start up. MPQ5073 also provides programmable soft start time, output discharge functions, OCP and thermal shutdown features.

The max load at the output (source) is current limited. This is accomplished by utilizing a sense FET topology.

The magnitude of the current limit is controlled by an external resistor from the ILIM pin to ground.

An internal charge pump drives the gate of the power device, allowing a very low on-resistance DMOS power FET of just  $50m\Omega$ .

Tiny QFN-12 (2mmx2mm) of MPQ5073 is available in space saving package.

## **FEATURES**

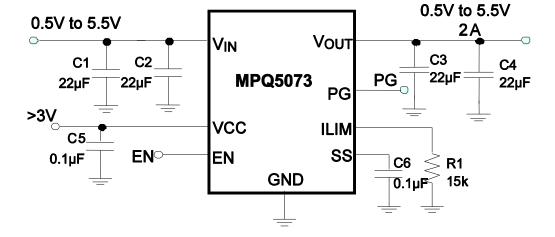
- Guaranteed Industrial/Automotive Temp Range Limits
- Large V<sub>IN</sub> Range from 0.5V to 5.5V
- <5µA Shutdown Current</li>
- Integrated 50mΩ Low R<sub>DSON</sub> FETs
- Typical 2A Load Current Range
- Push Pull PG Indicator
- Adjustable Start Up Slew Rate
- Output discharge function
- <200ns Short-Circuitry Response Protection</li>
- Thermal Protection
- Small QFN-12 (2mmx2mm) Package for Space Saving
- Available in AEC-Q100 Qualified Grade

## **APPLICATIONS**

- Notebook and Tablet Computers
- Portable Devices
- Solid State Drivers
- Handheld Devices

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## TYPICAL APPLICATION



## ORDERING INFORMATION

Part Number*	Package	Top Marking	
MPQ5073GG	QFN-12 (2mmx2mm)	See Below	
MPQ5073GG-AEC1	QFIN-12 (ZIIIIIXZIIIII)	See below	

<sup>\*</sup> For Tape & Reel, add suffix -Z (e.g. MPQ5073GG-Z);

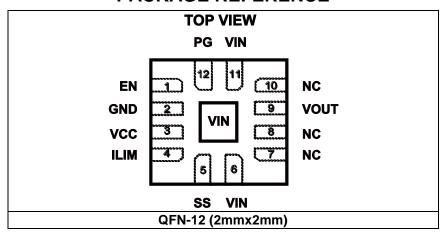
## **TOP MARKING**

CBY

CB: product code of MPQ5073GG & MPQ5073GG-AEC1;

Y: year code; LLL: lot number;

## **PACKAGE REFERENCE**



ABSOLUTE MAXIMUM RATINGS (1)
V <sub>IN</sub>
V <sub>CC</sub>
V <sub>OUT</sub>
EN, SS, ILIM0.3V to Vcc+0.3 V
Junction Temperature 150°C
Lead Temperature260°C
Continuous Power Dissipation (2)
QFN-12 (2mmx2mm)1.6W
Recommended Operating Conditions (3)
Supply Voltage V <sub>IN</sub> 0.5V to 5.5V
Supply Voltage V <sub>CC</sub>
Output Voltage V <sub>OUT</sub> 0.5V to 5.5V
Operating Junction Temp40°C to +125°C

Thermal Resistanc	$\mathbf{e}^{(4)}$ $\boldsymbol{\theta}_{JA}$	$oldsymbol{ heta}_{JC}$	
QFN-12 (2mmx2mm)	80	16	°C/W

#### Notes:

- Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature  $T_J$  (MAX), the junction-to-ambient thermal resistance  $\theta_{JA}$ , and the ambient temperature  $T_A$ . The maximum allowable continuous power dissipation at any ambient temperature is calculated by  $P_D$  (MAX) = ( $T_J$  (MAX)- $T_A$ )/ $\theta_{JA}$ . Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the device will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB.



## **ELECTRICAL CHARACTERISTICS**

 $V_{IN}$  = 3.6V,  $V_{CC}$  = 3.6V,  $T_J$  =-40°C to +125°C, typical values are  $T_J$  = 25°C, unless otherwise noted.

Parameters	Symbol	Condition	Min	Тур	Max	Units
Input and Supply Voltage Range						
Input Voltage	V <sub>IN</sub>		0.5		5.5	V
Supply Voltage	V <sub>CC</sub>		3		5.5	V
Supply Current						
Off State Leakage Current	loff	V <sub>IN</sub> =5V, EN=0, T <sub>J</sub> = 25°C			1	μΑ
Off State Leakage Current		V <sub>IN</sub> =5V, EN=0, -40°C < T <sub>J</sub> <125°C			5	μΑ
		V <sub>CC</sub> =5V, EN=0		0.1	1	
V <sub>CC</sub> Standby Current	I <sub>STBY</sub>	V <sub>CC</sub> =5V, Enable, No Load T <sub>J</sub> = 25°C		180	230	μА
<b></b>		V <sub>CC</sub> =5V, Enable, No Load, -40°C < T <sub>J</sub> <125°C		180	250	
Power FET		,			l.	1
		V <sub>CC</sub> =5.0V, T <sub>J</sub> =25°C		50	70	
ON Pasistanas	Ь	V <sub>CC</sub> =5.0V, -40°C <t<sub>J&lt;125°C</t<sub>		50	80	mΩ
ON Resistance	R <sub>DSON</sub>	V <sub>CC</sub> =3.3V, T <sub>J</sub> =25°C		60	80	
		V <sub>CC</sub> =3.3V, -40°C <t<sub>J&lt;125°C</t<sub>		60	90	
Thermal Shutdown and Recovery (5)						
Shutdown Temperature	T <sub>STD</sub>			150		°C
Hysteresis	T <sub>HYS</sub>			30		°C
Under Voltage Protection						
V <sub>CC</sub> Under Voltage Lockout Threshold	V <sub>CC_UVLO</sub>	UVLO Rising Threshold		2.6	2.95	V
UVLO Hysteresis	V <sub>UVLO_HYS</sub>			200		mV
Soft Start	•					
SS pull-up Current	I <sub>SS</sub>	Fixed Slew Rate	4	11	17	μA
Enable						
EN Rising Threshold	V <sub>ENH</sub>		1.3	1.5	1.7	V
EN Hysteresis	V <sub>EN_hys</sub>			200		mV
ILIM		1				1
Current Limit	I <sub>OUT</sub>	R <sub>LIM</sub> =24kΩ. Ramp I <sub>OUT</sub> Record Peak Current Limit Value		1.4		А
Discharge Resistance	•	•				
Discharge Resistance	R <sub>DIS</sub>			200		Ω
PG		1			1	1
Power Good Rising Threshold V <sub>PG_F</sub>		Voltage Gap between V <sub>OUT</sub> and V <sub>IN</sub>	140	280	450	mV
Power Good Threshold	V <sub>PG_HYS</sub>			60		mV
Power Good Delay	T <sub>PG_D</sub>			50		μs
Power Good High	V <sub>PG_H</sub>	V <sub>CC</sub> =3.3V	3.2	-	1	V
Power Good Low	V <sub>PG_L</sub>	Sink 1mA			0.3	V

### Notes:

<sup>5)</sup> Guarantee by design.



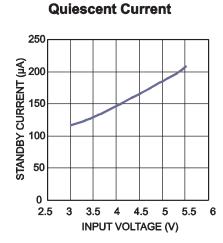
# **PIN FUNCTIONS**

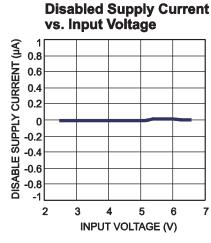
QFN-12 (2mmx2mm) Pin #	Name	Description
1	EN	Enable Input. Pulling this pin below the specified threshold shuts the chip down.
2	GND	Ground.
3	VCC	Supply Voltage to the Control Circuitry.
4	ILIM	Output Current Limit Configure. Place a resistor to ground to set the overload current limit level.
5	SS	Soft Start Pin. An external capacitor connected to this pin sets the slew rate of the output voltage soft start period.
6, 11, Exposed pad	VIN	Input Power Supply.
9	VOUT	Output to the Load.
12	PG	Power Good Pin. Push-Pull output.
7,8,10	NC	NC Pin, Suggest connecting them with VOUT to improve the thermal performance.

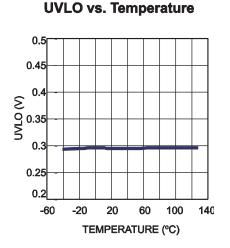


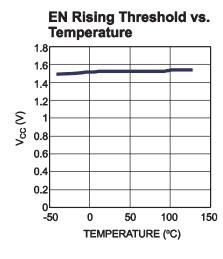
## TYPICAL PERFORMANCE CHARACTERISTICS

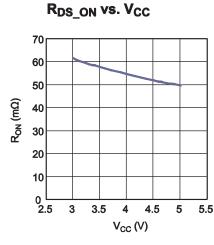
 $V_{IN} = 3.6V$ ,  $V_{CC} = 3.6V$ ,  $R_{ILIM} = 13k\Omega$ ,  $T_A = 25$ °C, unless otherwise noted.

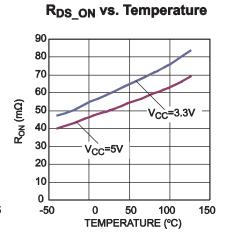


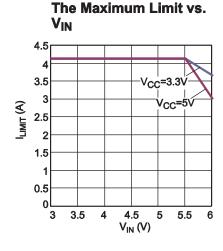


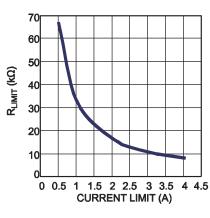




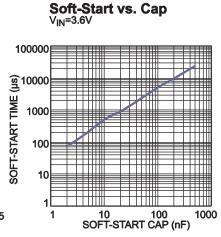








Current Limit vs. RLIMIT

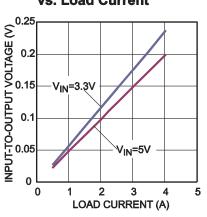




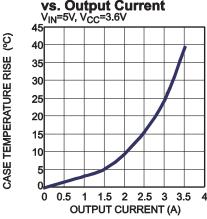
# TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 $V_{IN} = 3.6V$ ,  $V_{CC} = 3.6V$ ,  $R_{ILIM} = 13k\Omega$ ,  $T_A = 25^{\circ}C$ , unless otherwise noted.

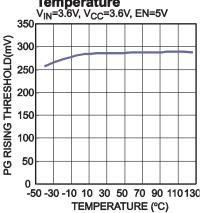




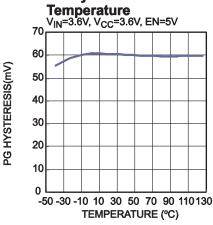
**Case Temperature Rise** vs. Output Current







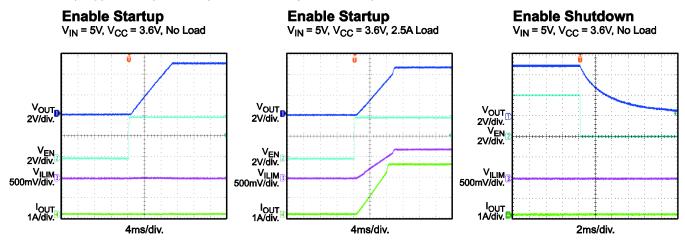
## PG Hysteresis vs.

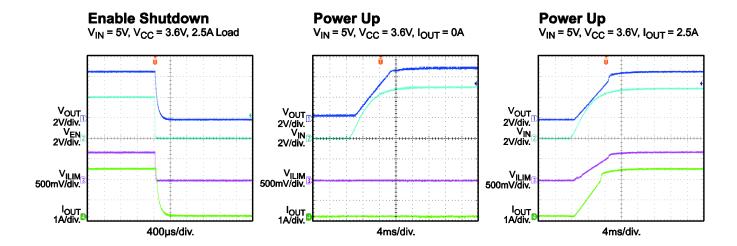


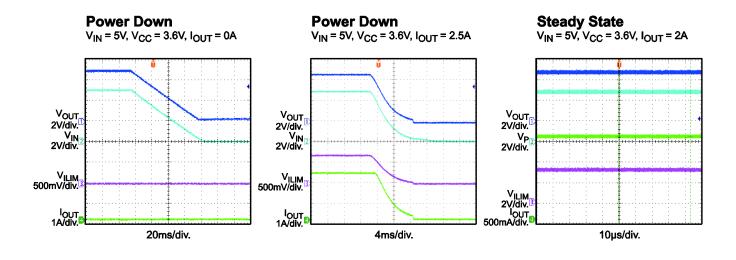


## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 $V_{IN} = 3.6V$ ,  $V_{CC} = 3.6V$ , EN=4V,  $R_{ILIM} = 13k\Omega$ ,  $T_A = 25$ °C, unless otherwise noted.



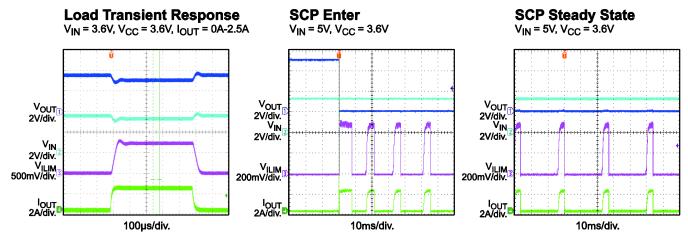




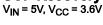


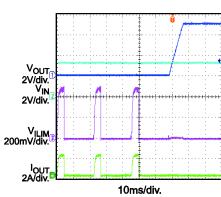
# TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 $V_{IN}$  = 3.6V,  $V_{CC}$  = 3.6V, EN=4V,  $R_{ILIM}$  = 13k $\Omega$ ,  $T_A$  = 25°C, unless otherwise noted.









### **OPERATION**

The MPQ5073 is designed to limit the in-rush current to the load when a circuit card is inserted into a live backplane power source; thereby limiting the backplane's voltage drop and the slew rate of the voltage to the load. It provides an integrated solution to monitor the input voltage, output voltage and output current to eliminate the need for an external current power MOSFET, and current sense device.

#### **Enable**

When input voltage is greater than the undervoltage lockout threshold (UVLO), typically 0.5V, MPQ5073 can be enabled by pulling EN pin to higher than 1.5V. Pulling down to ground will disable MPQ5073.

#### **Current Limit**

The MPQ5073 provides a constant current limit that can be programmed by an external resistor. Once the device reaches its current limit threshold, the internal circuit regulates the gate voltage to hold the current in the power FET constant. The typical response time is about 20µs and the output current may have a small overshoot during this time period.

The pre-set current limit value can be calculated by below equation:

$$I_{Limit} = (1 \div R_{ILIM}) \times S \tag{1}$$

S is the current sense ratio of MPQ5073, and this value is typically 33000 in  $V_{IN}$ =3.6V.

If the current limit block starts to regulate the output current, the power loss on power MOSFET will cause the IC temperature rise. If the junction temperature rose to high enough, it will trigger thermal shutdown. After thermal shutdown happened, it will disable the output until the over temperature fault remove. The over temperature threshold is 150°C and hysteresis is 30°C.

#### **Power-Good Function**

The PG pin is the push pull of a MOSFET that can be pulled high to  $V_{\rm CC}$ . The MOSFET turns on with the application of an input voltage so that the PG pin is pulled to GND. After the voltage gap between  $V_{\rm IN}$  and  $V_{\rm OUT}$  is smaller

than 280mV, the PG pin is pulled high after a 50µs delay. When the voltage gap is higher than 340mV, the PG pin will be pulled low.

#### **Short-Circuit Protection**

If the load current increases rapidly due to a short circuit, the current may exceed the current limit threshold by a lot before the control loop can respond. If the current reaches an internal secondary current limit level (about 7A), a fast turn-off circuit activates to turn off the power FET. This limits the peak current through the switch to limit the input voltage drop. The total short circuit response time is about 200ns. If fast off works, it will keep off the power FET for 80µs. After that time period, it will re-turn on power FET, if the part is still in short-circuit condition. MPQ5073 will reduce the current limit, and hold it until the part is so hot and thermal shutdown. After the short-circuit condition removed, the current limit will recover to the pre-set value automatically.

### **Output Discharge**

MPQ5073 has output discharge function. This function can discharge the Vo by internal pull down resistance when IC disabled and the load is very light.

### **Soft-Start**

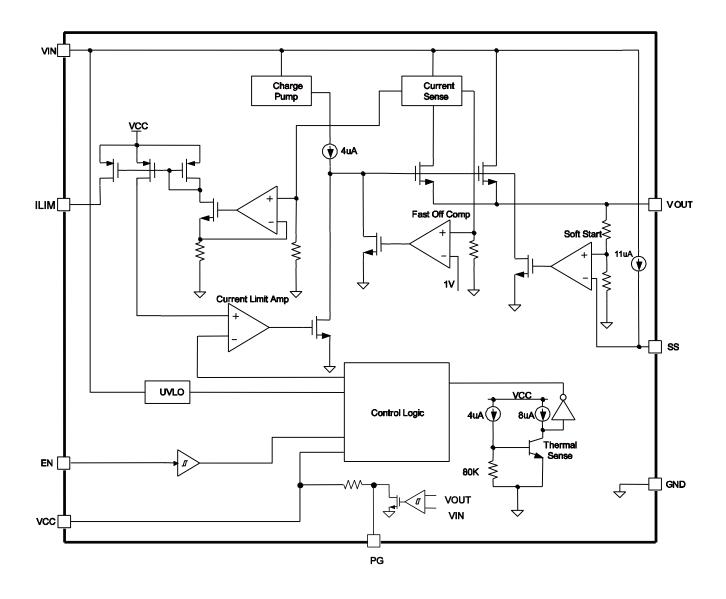
A capacitor connected to the SS pin determines the soft-start time. There is an internal  $11\mu$ A constant current source charge SS cap and ramps up the voltage on the SS pin. The output voltage rises at 5 times the slew rate to the SS voltage.

The soft-start time can be calculated by below equation:

$$T_{SS}(ms) = \frac{1}{5} \times \frac{V_{OUT}(V) \cdot C_{SS}(nF)}{I_{SS}(uA)}$$
 (2)

 $T_{SS}$  is the soft-start time,  $I_{SS}$  is internal 11µA constant current,  $C_{SS}$  is external soft-start cap.

The suggestion minimum SS cap should be bigger than 4.7nF. If the SS pin is floated or SS cap is too small, the  $V_{\text{OUT}}$  rising time will be just limited by power MOS charge time.



**Figure 1: Functional Block Diagram** 

## APPLICATION INFORMATION

#### **ILIM Resistor Selection**

The current limit value can be set by ILIM resistor. The current limit can be gotten by equation (1).

The current limit threshold is suggested to 10% ~ 20% higher than maximum load current. For example, if the system's full load is 2A, set the current limit to 2.2A.

## **ILIM Capacitor Selection**

The internal advanced auto-zero comparator bring a high accuracy of current monitor. The auto-zero will also cause some little jitter on ILIM pin. To get a more stable ILIM, a small ceramic capacitor can be mounted between ILIM and ground. Suggested place an ILIM capacitor less than 1nF.

## **Soft Start Capacitor Selection**

There is an internal  $11\mu A$  constant current source charge SS cap and ramps up the voltage on the SS pin. The output voltage rises follow the slew rate of SS voltage.

If the inrush on output current reached the current limit during start up (like with large output cap or very large load), MPQ5073 will limit the output current and the same time, SS time will be increased (Fig 2 and Fig 3).

### **Component Selection**

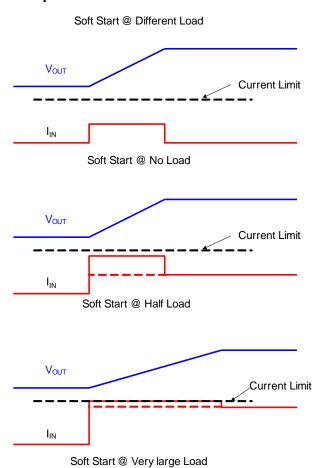
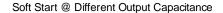
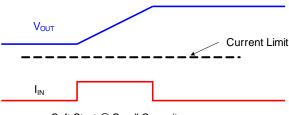
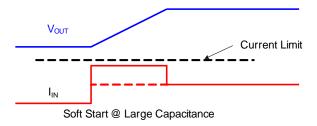


Figure 2: Soft Start Periods at different load





Soft Start @ Small Capacitance



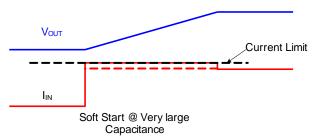


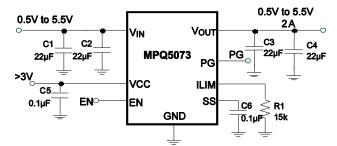
Figure 3: Soft Start Periods at different output capacitance

## **Design Example**

Some design example and are provided below. See Table 1 and Figure 4.

Table 1

(V)	Max Load Range (A)	Rlimit (kΩ)	SS cap (nF)	SS time (ms)
3.6	0.5	47	22	1
3.6	1	27.4	47	2.4
3.6	2	15	100	5.4



**Figure 4: Typical Application Schematic** 

## **Layout Guide**

PCB layout is very important to achieve stable operation. Please follow these guidelines and take below figure for reference. Place  $R_{\rm ILIM}$  close to ILIM pin, input cap close to  $V_{\rm CC}$  pin. Put enough vias around IC to achieve better thermal performance.

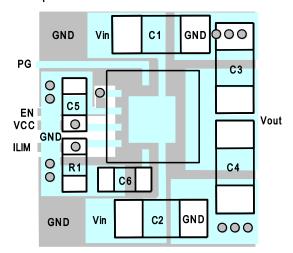
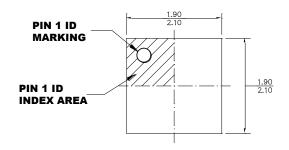
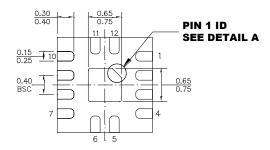


Figure 5: Recommended Layout

## **PACKAGE INFORMATION**

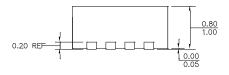
## QFN-12 (2mmx2mm)



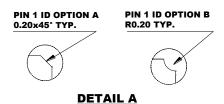


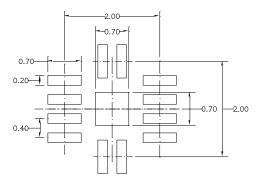
#### **TOP VIEW**

**BOTTOM VIEW** 



**SIDE VIEW** 





# NOTE:

- 1) ALL DIMENSIONS ARE IN MILLIMETERS.
- 2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
- 3) LEAD COPLANARITY SHALL BE 0.10 MILLIMETERS MAX.
- 4) JEDEC REFERENCE IS MO-229.
- 5) DRAWING IS NOT TO SCALE.

## RECOMMENDED LAND PATTERN

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