

## 8bit High-speed ADC

### PRODUCT DESCRIPTION

The MS5510 is a 8-bit, 20MSPS analog-to-digital converter (ADC) using semiflash architecture. The MS5510 operates in 5V supply and the typical power dissipation is only 130mW. The MS5510 includes an internal sample-and-hold circuit, parallel output with high-impedance mode, and internal reference resistor.

The semiflash architecture reduces power dissipation and die size compared with flash converter. By completing the conversion in a 2-step process, the number of comparator can be greatly reduced. The wait time of converted data is 2.5 clocks.

The MS5510 has two operation modes. Mode 1: use three internal reference resistors to connect with VDDA, which can generate a standard 2V full-scale conversion range. Only external jumper is needed to complete this option. Mode 2: by internal resistor section, a standard 4V full-scale conversion range is generated, which reduces the requirements for external reference or resistor. Differential linearity is 0.5LSB at 25°C and the maximum is 0.75LSB over the full operating temperature range. Typical dynamic specifications include a differential gain of 1% and differential phase of 0.7 degrees.

The operating temperature ranges from -20°C to 75°C.

### FEATURES

- Analog Signal Input Range : Mode 1: 2V (Max); Mode 2: 4V (Max)
- Resolution : 8bit
- INL:  $\pm 0.75$  LSB (25°C);  $\pm 1$  LSB (-20°C-75°C)
- DNL:  $\pm 0.5$  LSB (25°C);  $\pm 0.75$  LSB (-20°C-75°C)
- Fastest Conversion Rate : 20MSPS
- 5V Single Power Supply
- Low Power Dissipation : Mode 1 : 127.5mW ; Mode 2: 150mW



**SOP24**

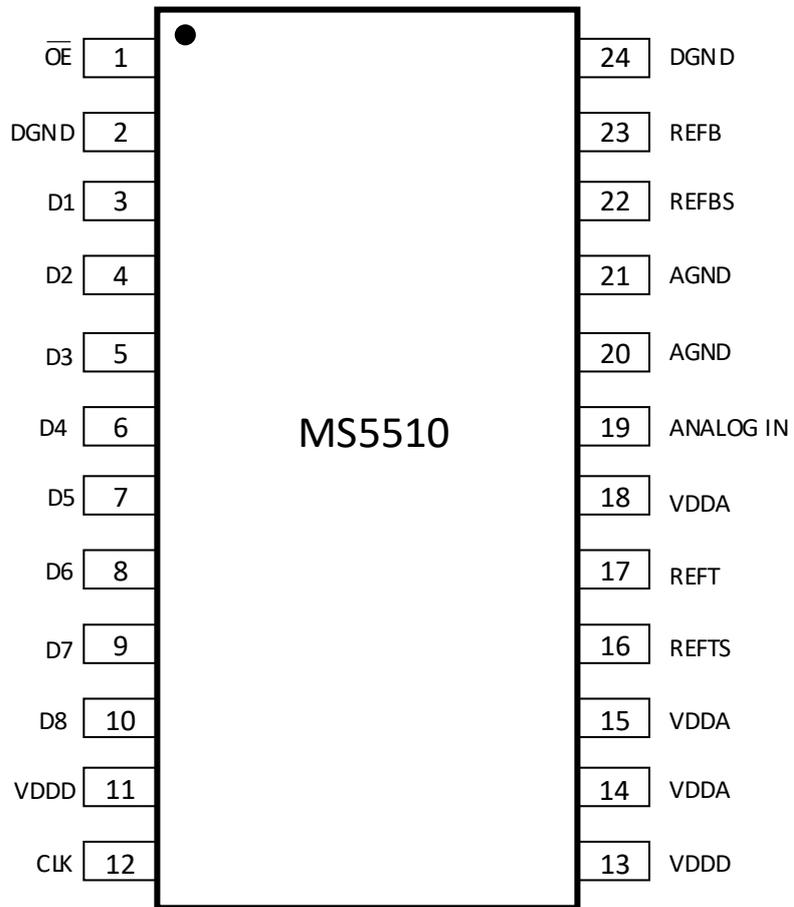
### APPLCATIONS

- DTV
- Multimedia Image Process
- Video Conference
- High-speed Data Conversion
- Quadrature Demodulator

### PRODUCT SPECIFICATION

Part Number	Package	Marking
MS5510	SOP24	MS5510

**PIN CONFIGURATION**



**PIN DESCRIPTION**

Pin	Name	Type	Description
AGND	20, 21	-	Analog Ground
ANALOG IN	19	I	Analog Input
CLK	12	I	Clock Input
DGND	2,24	-	Digital Ground
D1-D8	3-10	O	Digital Data Output. D1=LSB, D8=MSB
$\overline{\text{OE}}$	1	I	Output Enable. When $\overline{\text{OE}}$ is 0, output data; When $\overline{\text{OE}}$ is 1, output high-impedance
VDDA	14,15,18	-	Analog Power Supply
VDDD	11,13	-	Digital Power Supply
REFB	23	I	Low Reference Voltage Input
REFBS	22	-	Low Reference Voltage. When using mode 1, internal voltage divider generates a rated 2V reference and REFBS is shorted to REFB. When using mode 2, REFBS is connected to ground.
REFT	17	I	High Reference Voltage Input
REFTS	16	-	High Reference Voltage. When using mode 1, internal voltage divider generates a rated 2V reference and REFTS is shorted to REFT. When using mode 2, REFTS is connected to VDDA.



**ABSOLUTE MAXIMUM RATINGS**

Any exceeding absolute maximum rating application causes permanent damage to device. Because long-time absolute operation state affects device reliability. Absolute ratings just conclude from a series of extreme tests. It doesn't represent chip can operate normally in these extreme conditions.

Parameter	Symbol	Ratings	Unit
Power Supply	$V_{DDA}, V_{DDD}$	7V	V
Reference Voltage Input	$V_{REFT}, V_{REFB}$	AGND ~ $V_{DDA}$	V
Analog Input Voltage	$V_I(ANLG)$	AGND ~ $V_{DDA}$	V
Digital Input Voltage	$V_I(DGTL)$	DGND ~ $V_{DDD}$	V
Digital Output Voltage	$V_O(DGTL)$	DGND ~ $V_{DDD}$	V
Operating Temperature	$T_A$	-20 ~ 75	°C
Storage Temperature	$T_{stg}$	-55 ~ 150	°C

**RECOMMENDED OPERATING CONDITIONS**

Parameter	Condition	Min	Typ	Max	Unit
Power Supply	$V_{DDA}$ -AGND	4.75	5	5.25	V
	$V_{DDD}$ -AGND	4.75	5	5.25	
	AGND-DGND	-100	0	100	mV
Reference Input Voltage (Top), $V_{ref(T)}$	Mode 2	$V_{REFB}+2$		4	V
Reference Input Voltage (Bottom), $V_{ref(B)}$	Mode 2	0		$V_{REFT}-4$	V
Analog Input Voltage, $V_I(ANLG)$		$V_{REFB}$		$V_{REFT}$	V
High-level Input Voltage, $V_{IH}$		4			V
Low-level Input Voltage, $V_{IL}$				1	V
Pulse Width, Clock High-level, $t_{w(H)}$		25			ns
Pulse Width, Clock Low-level, $t_{w(L)}$		25			ns

**ELECTRICAL CHARACTERISTICS**

 Unless otherwise noted,  $V_{DD}=5V$ ,  $V_{REFT}=2.5V$ ,  $V_{REFB}=0.5V$ ,  $f_{(CLK)}=20MHz$ ,  $T_A=25^{\circ}C$ .

**Digital I/O**

Parameter	Symbol	Condition	Min	Typ	Max	Unit
High-level Input Current	$I_{IH}$	$V_{DD}=MAX, V_{IH}=V_{DD}$			5	$\mu A$
Low-level Input Current	$I_{IL}$	$V_{DD}=MAX, V_{IL}=0$			5	
High-level Output Current	$I_{OH}$	$\overline{OE}=GND, V_{DD}=MIN, V_{OH}=V_{DD}-0.5V$	-1.5			mA
Low-level Output Current	$I_{OL}$	$\overline{OE}=GND, V_{DD}=MIN, V_{OL}=0.4V$	2.5			
High-level High-impedance Output Leakage Current	$I_{OZH}$	$\overline{OE}=V_{DD}, V_{DD}=MAX, V_{OH}=V_{DD}$			16	$\mu A$
Low-level High-impedance Output Leakage Current	$I_{OZL}$	$\overline{OE}=V_{DD}, V_{DD}=MIN, V_{OL}=0$			16	

**Power Dissipation**

Parameter	Symbol	Condition	Min	Typ	Max	Unit	
Power Supply Current	$I_{DD}$	$f_{(CLK)}=20MHz$ , NTSC Slope-wave Input		18	27	mA	
Reference Voltage Current	$I_{ref}$	Mode 1	$V_{ref}=REFT-REFB=2V$	5.2	7.5	10.5	mA
		Mode 2	$V_{ref}=REFT-REFB=4V$	10.4	15	21	mA

**Static Performance**

Parameter	Symbol	Condition	Min	Typ	Max	Unit	
Self-bias (1)	REFB	REFB is shorted to REFB <sub>S</sub> .	0.57	0.61	0.65	V	
Self-bias (2)	REFT-REFB	REFT is shorted to REFT <sub>S</sub>	1.9	2.02	2.15	V	
Self-bias (3)	REFT	REFB is connected to AGND. REFT is connected to REFT <sub>S</sub>	2.18	2.29	2.4	V	
Reference Voltage Resistor	$R_{ref}$	Between REFT and REFB	190	270	350	$\Omega$	
Analog Input Capacitance	$C_i$	$V_{I(ANLG)}=1.5V+0.07V_{rms}$		16		pF	
Integral Nonlinearity	INL	Mode 1	$f_{(CLK)}=20MHz$ , $T_A=25^{\circ}C$		$\pm 0.4$	$\pm 0.75$	
			$V_i=0.5V$ to $2.5V$ $T_A=-20^{\circ}C-75^{\circ}C$			$\pm 1$	
		Mode 2	$f_{(CLK)}=20MHz$ , $T_A=25^{\circ}C$		$\pm 0.4$	$\pm 0.7$	
			$V_i=0$ to $4V$ $T_A=-20^{\circ}C-75^{\circ}C$				$\pm 1$

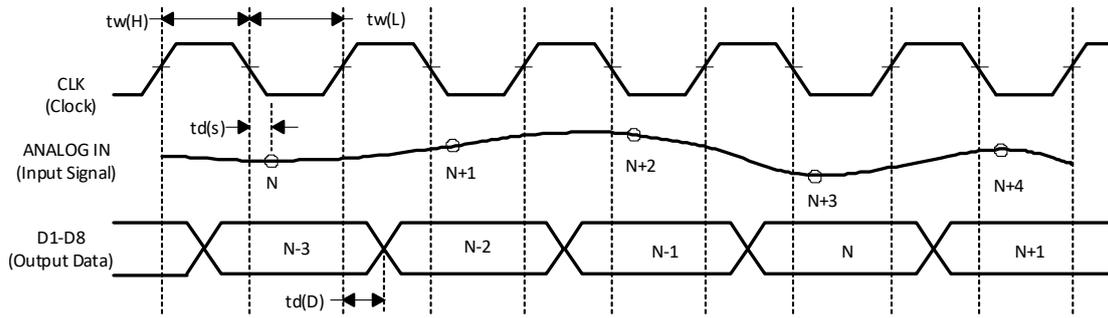
Parameter	Symbol	Condition		Min	Typ	Max	Unit
Differential Nonlinearity	DNL	Mode 1	$f_{(CLK)}=20\text{MHz}$ , $V_I=0.5\text{V to }2.5\text{V}$	$T_A=25^\circ\text{C}$	$\pm 0.3$	$\pm 0.5$	
				$T_A=-20^\circ\text{C}-75^\circ\text{C}$			$\pm 0.7$
		Mode 2	$f_{(CLK)}=20\text{MHz}$ , $V_I=0 \text{ to } 4\text{V}$	$T_A=25^\circ\text{C}$	$\pm 0.3$	$\pm 0.5$	
				$T_A=-20^\circ\text{C}-75^\circ\text{C}$			$\pm 0.7$
Zero-scale Error	$E_{ZS}$	Mode 1	$V_{ref}=REFT-REFB=2\text{V}$	-18	-43	-68	mV
		Mode 2	$V_{ref}=REFT-REFB=4\text{V}$	-36	-86	-136	mV
Full-scale Error	$E_{FS}$	Mode 1	$V_{ref}=REFT-REFB=2\text{V}$	-20	0	20	mV
		Mode 2	$V_{ref}=REFT-REFB=4\text{V}$	-40	0	40	mV

### Operating Characteristics

Unless otherwise noted,  $V_{DD}=5\text{V}$ ,  $V_{REFT}=2.5\text{V}$ ,  $V_{REFB}=0.5\text{V}$ ,  $f_{(CLK)}=20\text{MHz}$ ,  $T_A=25^\circ\text{C}$ .

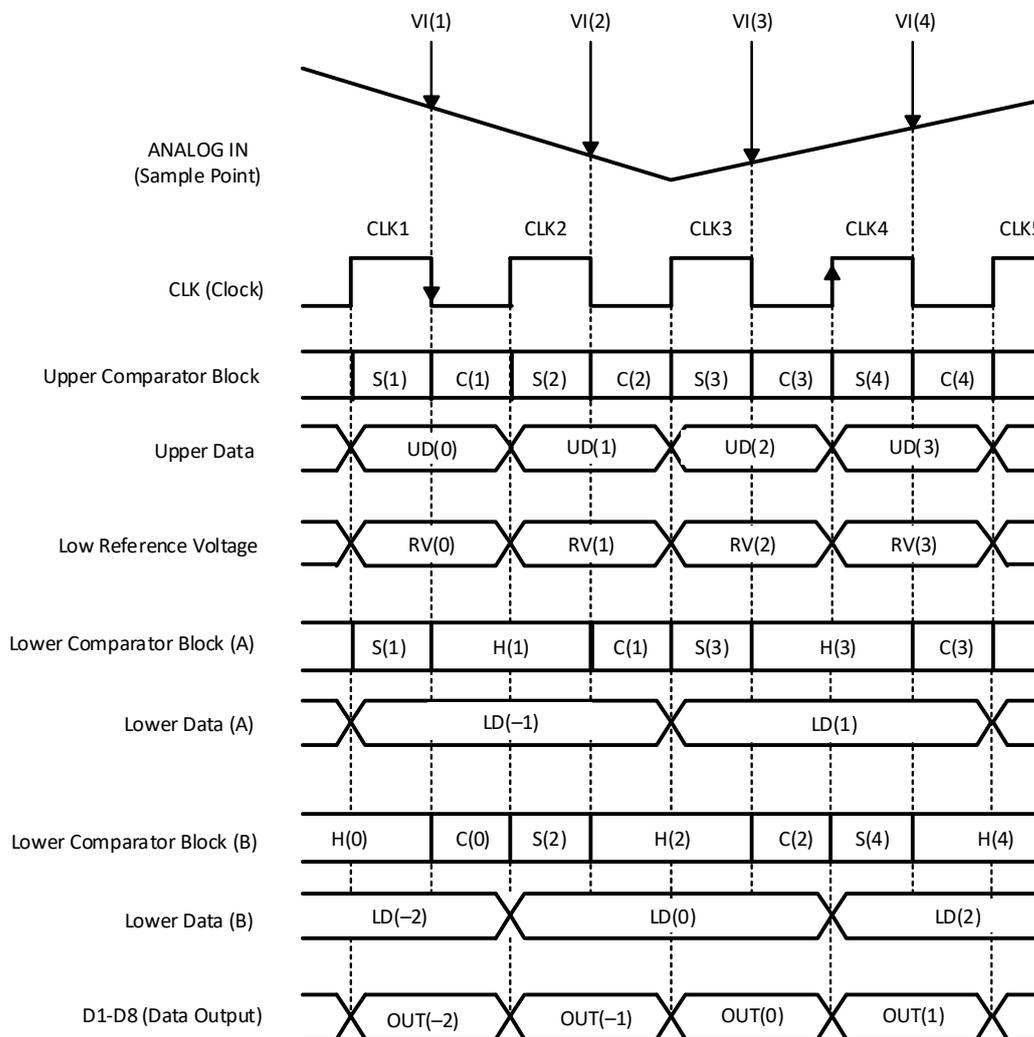
Parameter	Symbol	Condition		Min	Typ	Max	Unit
Maximum Conversion Ratio	$f_{conv}$	Mode 1	$f_i=1\text{kHz}$ $V_{I(ANLG)}=0.5-2.5\text{V}$			20	MSPS
		Mode 2	Slope-wave $V_{I(ANLG)}=0-4\text{V}$			20	MSPS
Analog Input Bandwidth	BW	@-1dB			14		MHz
Digital Output Delay	$t_{d(D)}$	$C_L \leq 10\text{pF}$			18	30	ns
Differential Gain		NTSC40 IRE Modulation Wave,			1%		
Differential Phase		$f_{conv}=14.3\text{MSPS}$			0.7		°
Sample Jitter Time	$t_{AJ}$				30		ps
Sample Delay Time	$t_{d(s)}$				4		ns
Enable Time. $\overline{OE}$ Falling Edge to Output Valid Data	$t_{en}$	$C_L=10\text{pF}$			5		ns
Disable Time. $\overline{OE}$ Rising Edge to Output High-impedance	$t_{dis}$	$C_L=10\text{pF}$			7		ns
Spurious Free Dynamic Range	SFDR	Input 1MHz	$T_A=25^\circ\text{C}$		45		dB
			Full Range		43		
		Input 3MHz	$T_A=25^\circ\text{C}$		45		
			Full Range		46		
		Input 6MHz	$T_A=25^\circ\text{C}$		43		
			Full Range		42		
		Input 10MHz	$T_A=25^\circ\text{C}$		39		
			Full Range		39		
Signal to Noise Ratio	SNR	$T_A=25^\circ\text{C}$			46		dB
		Full Range			44		

I/O Timing



**FUNCTION DESCRIPTION**

The MS5510 is a semiflash ADC with two lower comparator blocks (a comparator every four bits). As shown in following diagram, input voltage VI(1) is sampled on the falling edge of CLK1 and enter into the upper comparator block and the lower comparator block(A), S(1). The upper comparator block determines the upper data UD(1) on the rising edge of CLK2. At the same time, low reference voltage generates voltage RV(1) corresponding to the upper data. The lower comparator block (A) determines the lower data LD(1) on the rising edge of CLK3. UD(1) and LD(1) are combined and output as OUT(1) on the rising edge of CLK4. As described above, output data is delayed 2.5 clocks from the analog input voltage sampling point. Input voltage VI(2) is sampled on the falling edge of CLK2. UD(2) is determined on the rising edge of CLK3. LD(2) is determined on the rising edge of CLK4 by the lower comparator block(B). OUT(2) data is output on the rising edge of CLK5.

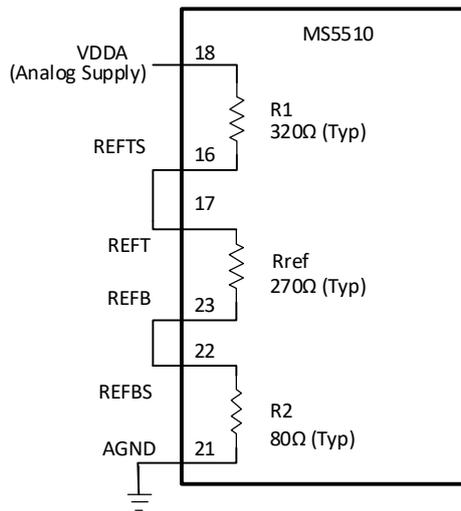


**Internal Reference**

**Operation Mode 1**

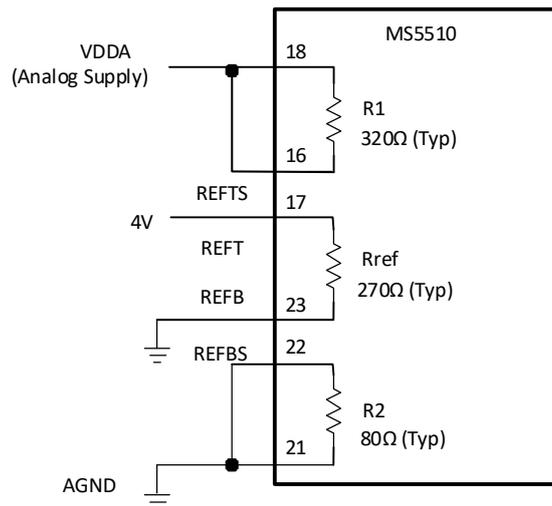
The MS5510 has three internal resistors to generate internal reference voltage. These resistors are connected to VDDA, REFTS, REFT, REFB, REFBS and AGND.

Using the internal reference is shown as follows. The connection provides the standard video 2V reference for the rated digital output.



**Operation Mode 2**

Analog input voltage range is 4V. REFT is connected with 4V, REFB is connected to ground and other connections are as follows. This connection provides the 4V reference for digital output with 0-4V analog input on ANALOG IN pin.



**Function Table**

Input Signal Voltage	Step	Digital Output Coding							
		MSB				LSB			
V <sub>ref</sub> (B)	255	0	0	0	0	0	0	0	0
.	.	.	.	.	.	.	.	.	.
.	.	.	.	.	.	.	.	.	.
.	128	0	1	1	1	1	1	1	1
.	127	1	0	0	0	0	0	0	0
.	.	.	.	.	.	.	.	.	.
.	.	.	.	.	.	.	.	.	.
V <sub>ref</sub> (T)	0	1	1	1	1	1	1	1	1

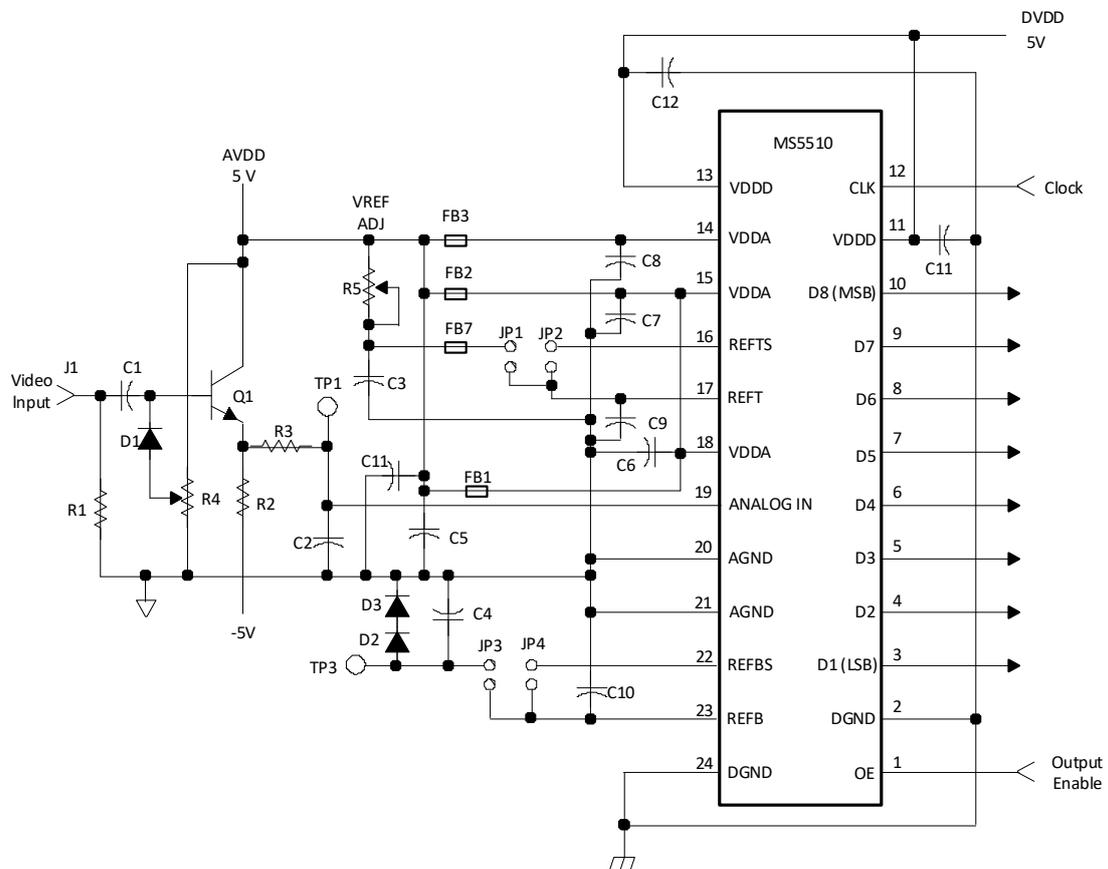
**APPLICATION INFORMATION**

The following contents are design recommendation items used with the MS5510.

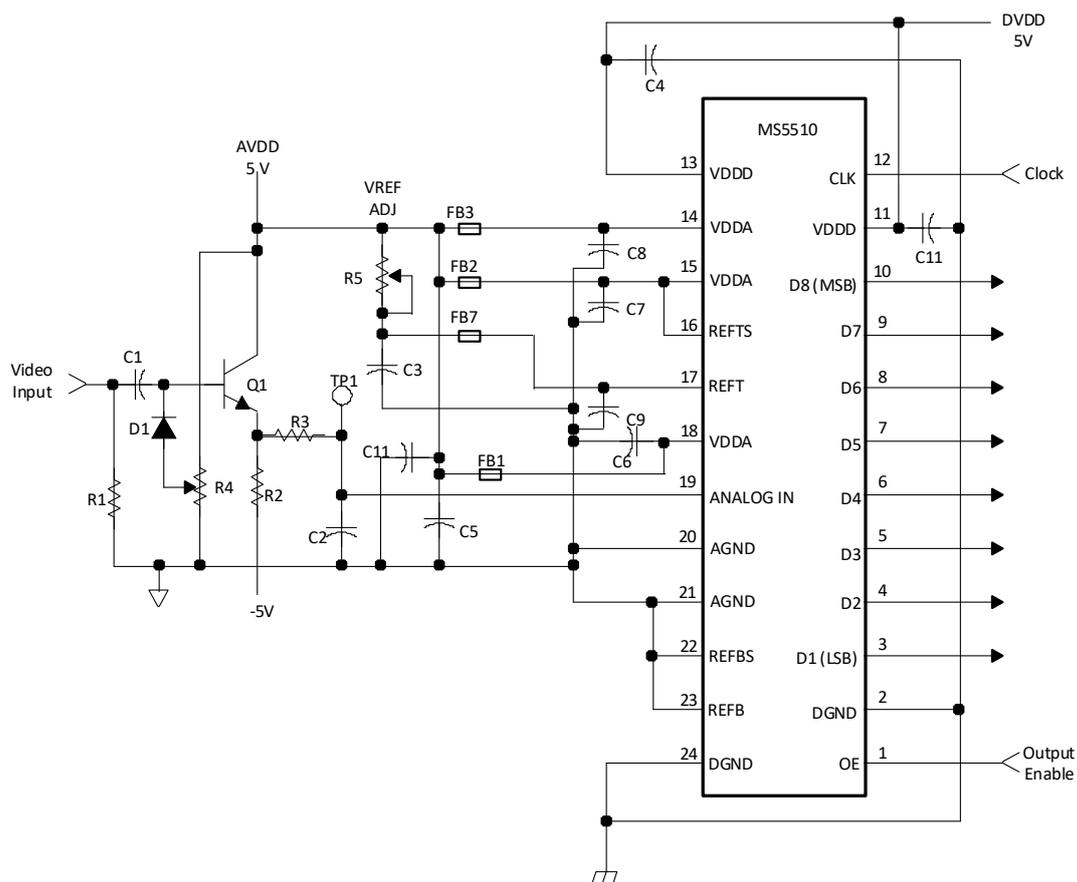
- External analog and digital circuit should be separated and shielded in order to reduce system noise.
- RF test board or PCB technology should be used during the evaluation and production process. And the test board should be copperized.
- Because AGND and DGND are not connected internally, these pins need to be connected externally. These ground traces should be connected with separate trace having better power bypass. In order to minimize the noise pickup, power line should use twisted-pair cables. An analog and digital ground plane should be used on PCB layout.
- VDDA to AGND and VDDD to DGND should be decoupled with 0.1μF capacitors respectively, which are placed as close as possible to these pins. Be careful to test in order to ensure a solid noise-free ground connection for the analog and digital ground.
- VDDA, AGND, and ANALOG IN should be separated from the high-frequency pins, CLK and D0–D7. When possible, AGND traces should be placed on both sides of the ANALOG IN on the PCB for shielding.
- When testing or using the MS5510, the resistance of the driving source connected to the analog input should be 10Ω or less within the analog frequency range.

**Evaluation and Test Diagram**

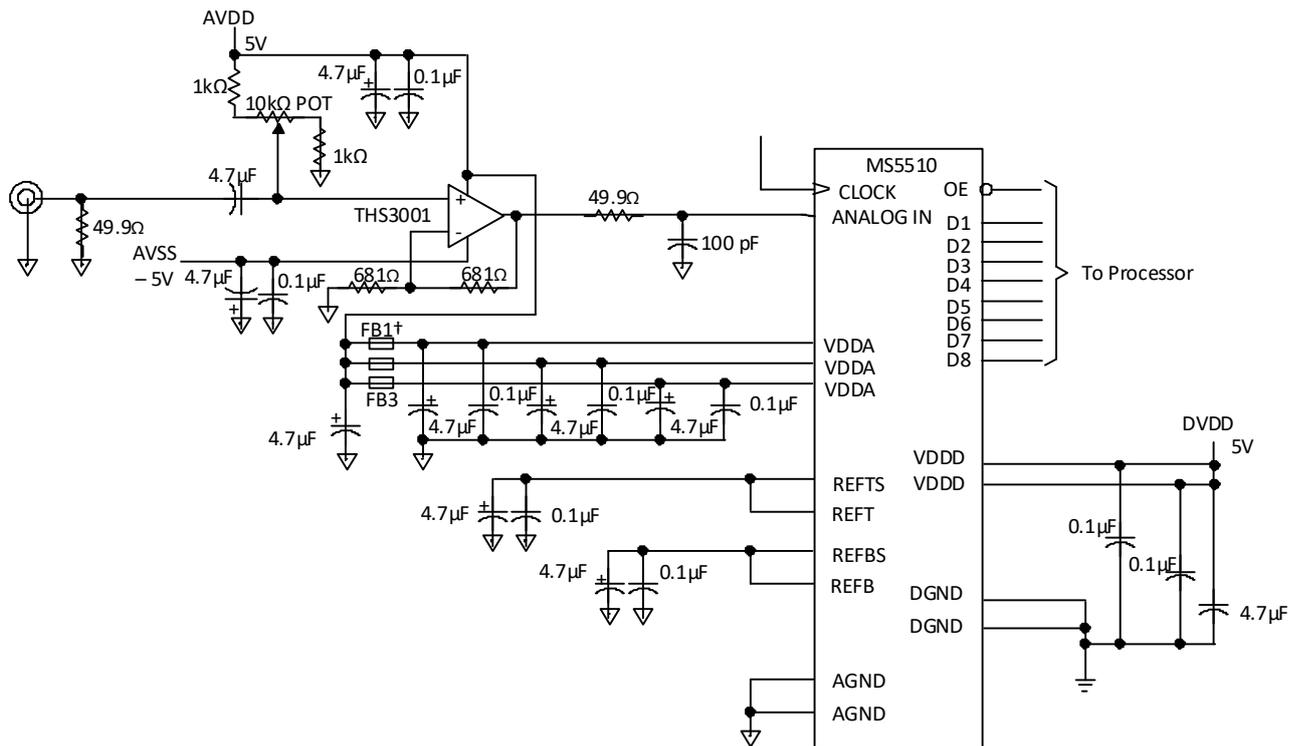
**Mode 1**



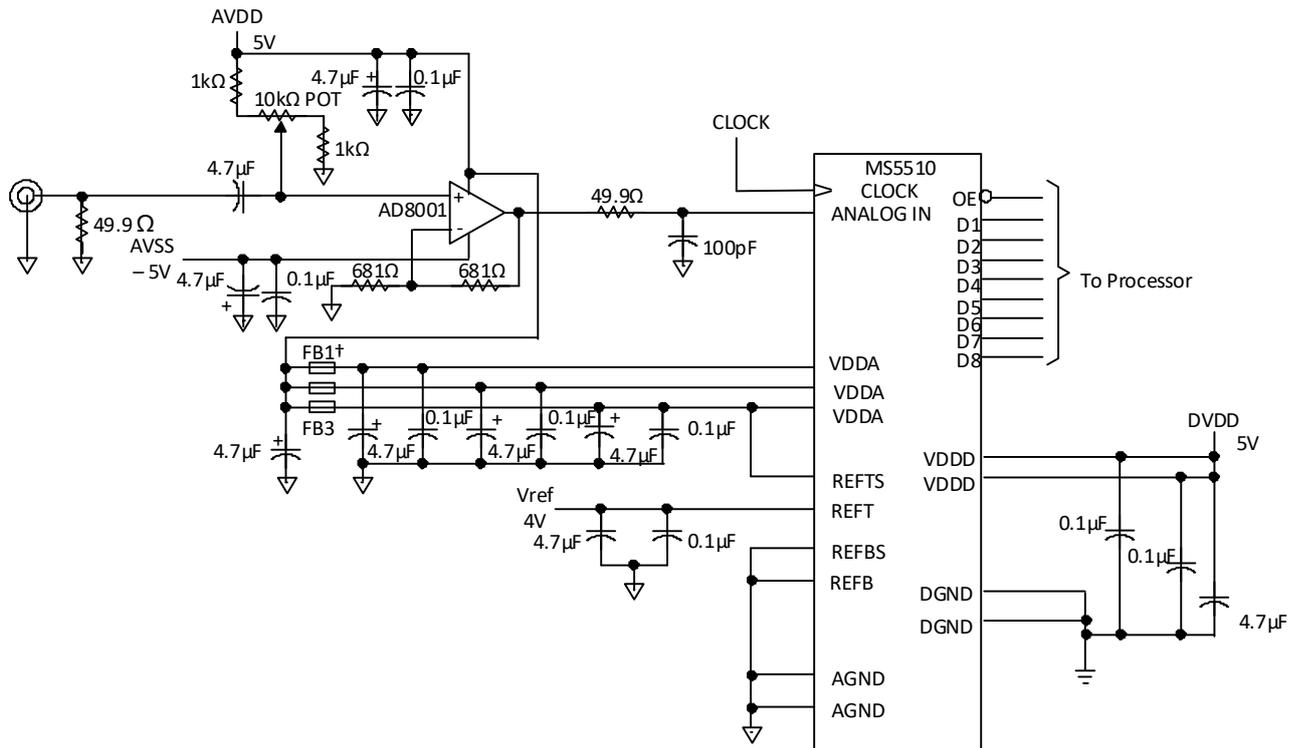
Element	Description
C1,C3-C4,C6-C12	0.1 $\mu$ f
C2	10pf
C5	47 $\mu$ f
FB1,FB2,FB3,FB7	Ferrite Ring
Q1	2N3414 or Equivalent Device
R1,R3	75 $\Omega$
R2	500 $\Omega$
R4	10k $\Omega$
R5	300 $\Omega$

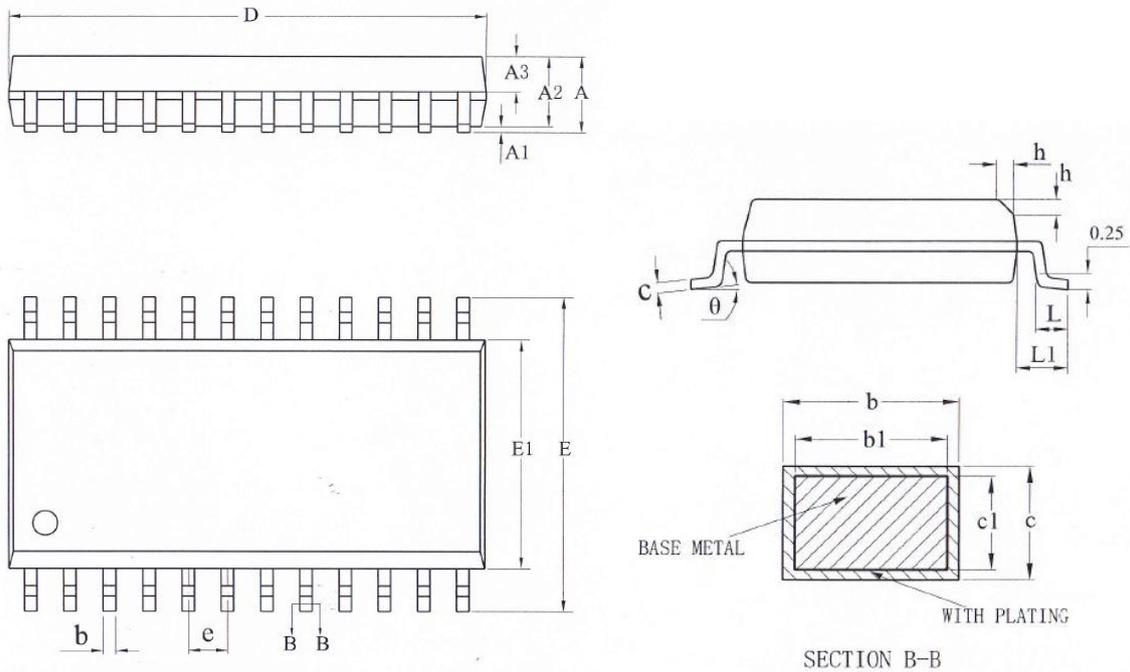
**Mode 2**


Element	Description
C1,C3-C4,C6-C11	0.1 $\mu$ f
C2	10pf
C5	47 $\mu$ f
FB1,FB2,FB3,FB7	Ferrite Ring
Q1	2N3414 or Equivalent Device
R1,R3	75 $\Omega$
R2	500 $\Omega$
R4	10k $\Omega$
R5	300 $\Omega$

**Application Diagram**
**Mode 1**


Mode 2



**PACKAGE OUTLINE DIMENSIONS**
**SOP24**


Symbol	Dimensions In Millimeters		
	Min	Typ	Max
A	2.36	2.54	2.64
A1	0.10	0.20	0.30
A2	2.26	2.30	2.35
A3	0.97	1.02	1.07
b	0.39	-	0.47
b1	0.38	0.41	0.44
c	0.25	-	0.29
c1	0.24	0.25	0.26
D	15.30	15.40	15.50
E	10.10	10.30	10.50
E1	7.40	7.50	7.60
e	1.27BSC		
L	0.70	-	1.00
L1	1.40REF		
h	0.25	-	0.75
θ	0	-	8°

**MARKING and PACKAGING SPECIFICATIONS**
**1. Marking Drawing Description**


Product Name : MS5510

Product Code : XXXXXX

**2. Marking Drawing Demand**

Laser printing, contents in the middle, font type Arial.

**3. Packaging Specifications**

Device	Package	Piece/Reel	Reel/Box	Piece /Box	Box/Carton	Piece/Carton
MS5510	SOP24	1000	1	1000	8	8000

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#### MOS CIRCUIT OPERATION PRECAUTIONS

Static electricity can be generated in many places. The following precautions can be taken to effectively prevent the damage of MOS circuit caused by electrostatic discharge:

1. The operator shall ground through the anti-static wristband.
2. The equipment shell must be grounded.
3. The tools used in the assembly process must be grounded.
4. Must use conductor packaging or anti-static materials packaging or transportation.



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